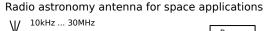
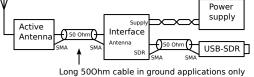
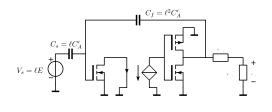
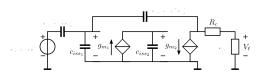
Design of a LF-HF Active Antenna in CMOS18 technology



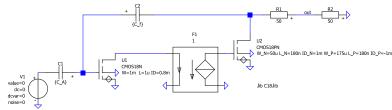


Design of a dual stage controller

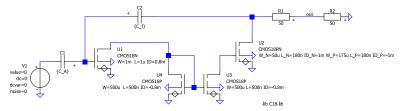




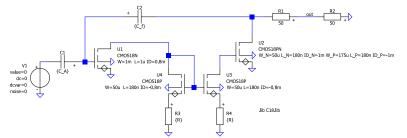
DualStage.py



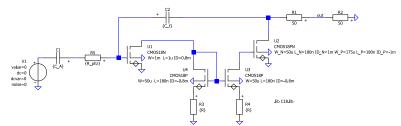
DualStageMirror.py



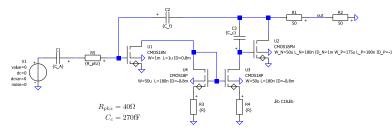
DualStageMirrorRes.py



DualStageMirrorResComp1.py



DualStageMirrorResComp2.py



| Output 1dB compression level: 0dBm in 500hm Antenna gain (-3dB: 10kHz-30MHz) | rotected CMOS technology | | | | |
|--|--|--|--|--|--|
| | | | | | |
| <i>C</i> . | | | | | |
| $L = -\frac{2g_{m_1}g_{m_2}R_c \frac{c_f}{C_f + C_s + c_{iss_1}}}{2g_{m_1}g_{m_2}R_c \frac{c_f}{C_f + C_s + c_{iss_1}}}$ | PZ analysis | | | | |
| $L = -\frac{2g_{m_1}g_{m_2}R_c \frac{c_f}{C_f + C_s + c_{iss_1}}}{sc_{iss_2} \left(1 + s2R_c \frac{c_f(C_s + c_{iss_1})}{C_f + C_s + c_{iss_1}}\right)}$ | PZ analysis results | | | | |
| | Gain type: gain | | | | |
| $q_{m_1}q_{m_2}$ | DC gain = -1.000 | | | | |
| $LP_2 = \frac{g_{m_1}g_{m_2}}{c_{iss_2}(C_s + c_{iss_1})}$ | | | | | |
| | pole Re [Hz] Im [Hz] Mag [Hz] Q p1 -3.513e+8 -1.207e+9 1.257e+9 1.789 | | | | |
| $g_{m_1} = 15.36 \text{m}$ | p2 -3.513e+8 1.207e+9 1.257e+9 1.789 | | | | |
| $g_{m_2} = 23.62 \mathrm{m}$ | zero Re [Hz] Im [Hz] Mag [Hz] Q | | | | |
| 0 | z1 2.176e+9 2.176e+9 | | | | |
| $C_s = 5 \mathrm{p}$ | z ₂ -3.978e+9 3.978e+9 | | | | |
| $c_{iss_1} = 7.568 \mathrm{p}$ | PZ analysis results | | | | |
| $c_{iss_2} = 0.4386 \text{p}$ | Gain type: loopgain | | | | |
| $B_f = \frac{1}{2\pi} \sqrt{LP_2} = 1.29 \text{GHz}$ | DC gain = oo | | | | |
| · 2/ | pole Re [Hz] Im [Hz] Mag [Hz] Q | | | | |
| $p_1 = 0, \ p_2 = -\frac{C_f + C_s + c_{iss_1}}{4\pi B_o C_f (C_s + c_{iss_1})} = -763 \text{MHz}$ | p1 0 0 | | | | |
| $4\pi R_c C_f \left(C_s + c_{iss_1} \right)$ | p ₂ -1.038e+9 1.038e+9 | | | | |
| ours of solos (abs) increased as a result of solo eslibili | zero Re [Hz] Im [Hz] Mag [Hz] Q | | | | |
| sum of poles (abs) increased as a result of pole-splitti in the second stage (first stage is shorted) | 5 21 3.313ers 3.313ers | | | | |
| in the second stage (iffst stage is shorted) | Z ₂ 4.979e+10 4.979e+10 | | | | |

Current mirror requirements

- Low noise:
 Transconductance low compared to input stage
 Low cut-off frequency (flicker noise)

- No dominant pole in loop gain
 Current mirror introduces a pole at half the cut-off frequency
 High cut-off frequency may conflict with low flicker noise

| | | PZ analysis | | | | | |
|---|----------------|-------------|-----------|----------|--------|--|--|
| F | Z ana | alysis resu | ults | | | | |
| G | ain ty | be: gain | | | | | |
| D | C gain = - | 0.9934 | | | | | |
| | pole | Re [Hz] | Im [Hz] | Mag [Hz] | Q | | |
| | P1 | -3.829e+7 | -5.721e+8 | 5.734e+8 | 7.488 | | |
| | p ₂ | -3.829e+7 | 5.721e+8 | 5.734e+8 | 7.488 | | |
| | P3 | -1.399e+9 | | 1.399e+9 | | | |
| | zero | Re [Hz] | Im [Hz] | Mag [Hz] | Q | | |
| | z1 | 1.113e+9 | | 1.113e+9 | | | |
| | Z2 | -8.226e+8 | -1.190e+9 | 1.447e+9 | 0.8793 | | |
| | z3 | -8.226e+8 | 1.190e+9 | 1.447e+9 | 0.8793 | | |

PZ analysis results

Gain type: loopgain

| pole | Re [Hz] | lm [Hz] | Mag [Hz] | Ş | MFM bandwidth | |
|----------------|-----------|---------|-----------|---|-------------------|--|
| P1 | -6.324e+6 | | 6.324e+6 | | B1 = 963MHz | |
| p ₂ | -4.850e+8 | | 4.850e+8 | | $B_{2} = 684 MHz$ | |
| p ₃ | -9.838e+8 | | 9.838e+8 | | $B_3 = 772 MHz$ | |
| zero | Re [Hz] | Im [Hz] | Mag [Hz] | Q | | |
| z1 | 1.106e+10 | | 1.106e+10 | | | |
| 75 | 4 979e+18 | | 4 979e+10 | | | |

Orthogonolize

Separate the function (inverting unity-gain current amplifier) from its noise performance (low transconductance)

Fix the transadmittance with negative feedback

| PZ analysis | | | | | | |
|----------------|--------------|-----------|-----------|--------|--|--|
| PZ ana | alysis resul | lts | | | | |
| Gain ty | pe: gain | | | | | |
| DC gain = - | 0.9930 | | | | | |
| pole | Re [Hz] | Im [Hz] | Mag [Hz] | Q | | |
| P1 | -7.126e+7 | -9.490e+8 | 9.517e+8 | 6.677 | | |
| p ₂ | -7.126e+7 | 9.490e+8 | 9.517e+8 | 6.677 | | |
| P3 | -2.320e+9 | | 2.320e+9 | | | |
| p4 | -3.258e+10 | | 3.258e+10 | | | |
| zero | Re [Hz] | Im [Hz] | Mag [Hz] | Q | | |
| z ₁ | 1.674e+9 | | 1.674e+9 | | | |
| Z2 | -1.533e+9 | -1.996e+9 | 2.516e+9 | 0.8209 | | |
| Z3 | -1.533e+9 | 1.996e+9 | 2.516e+9 | 0.8209 | | |
| z4 | -3.261e+10 | | 3.261e+10 | | | |

Frequency compensation

Second order system Influence of third pole cannot be ignored Compensation with phantom zero at the source introduces a fourth pole (resistor breaks loop of capacitors) Combination of pole-splitting and phantom-zero compensation

