#### **Structured Electronic Design**

Operational Amplifiers: modeling

Anton J.M. Montagne

Macro models

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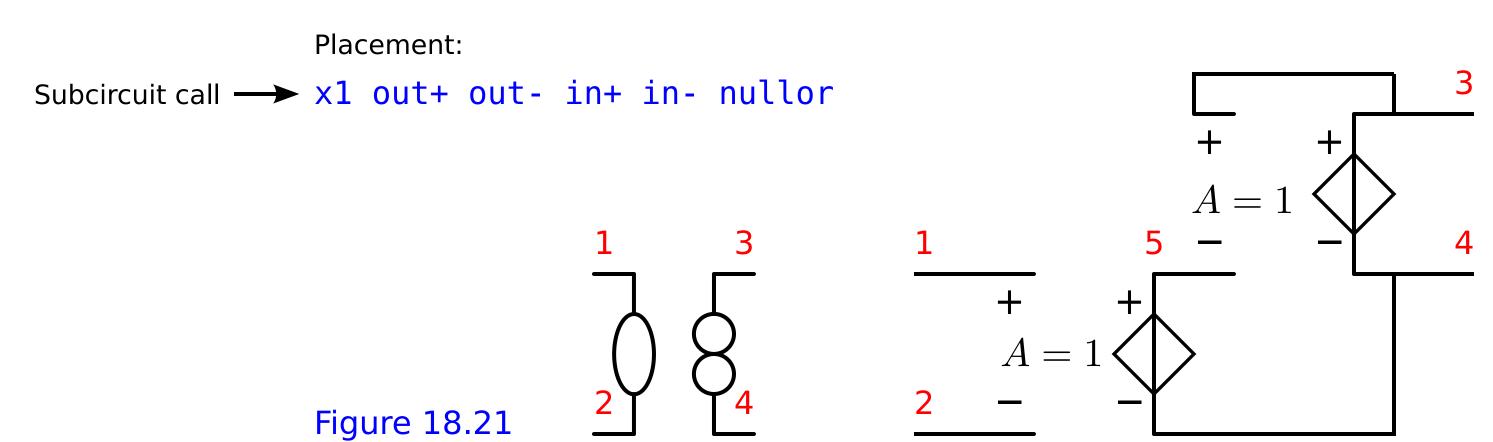
Spice nullor model

## Spice nullor model

Placement:

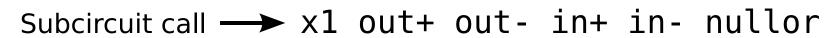
Subcircuit call —>

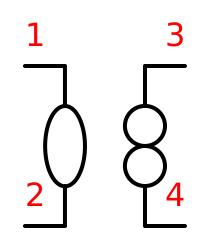
### Spice nullor model



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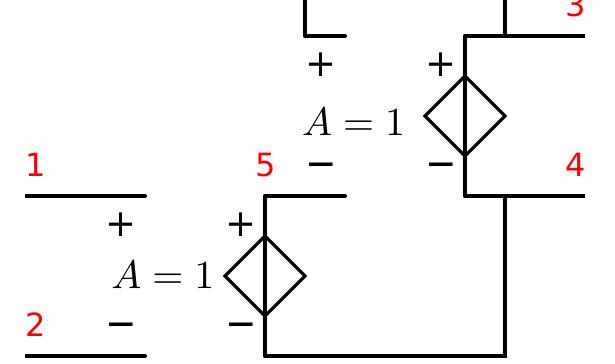


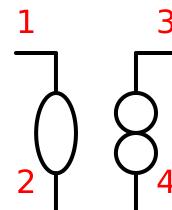
Figure 18.21

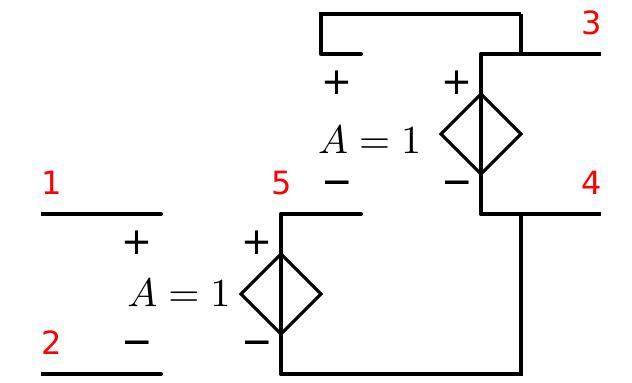
Definition:

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Subcircuit call  $\longrightarrow$  x1 out+ out- in+ in- nullor





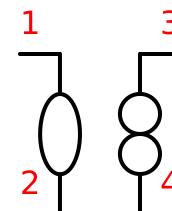
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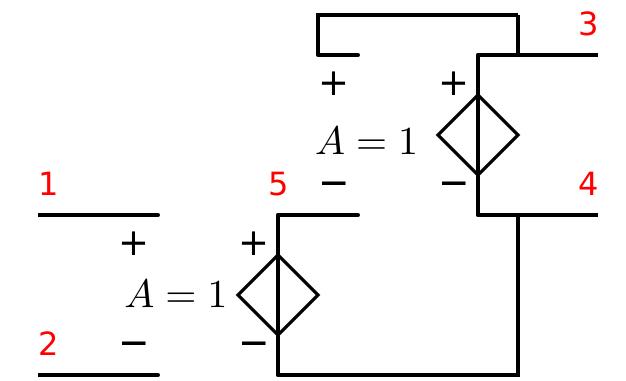
Figure 18.21

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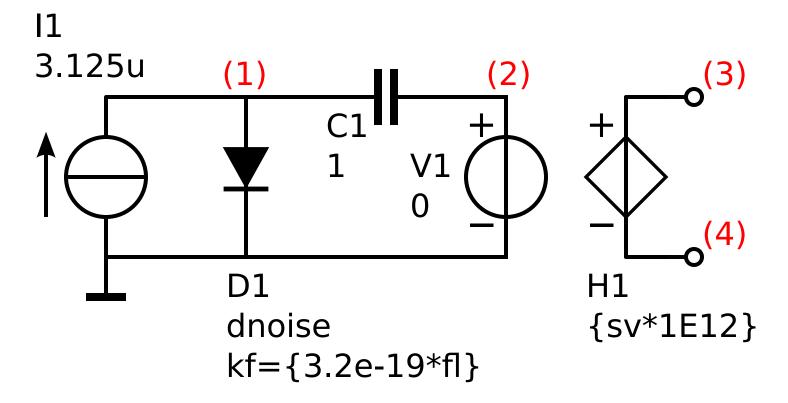


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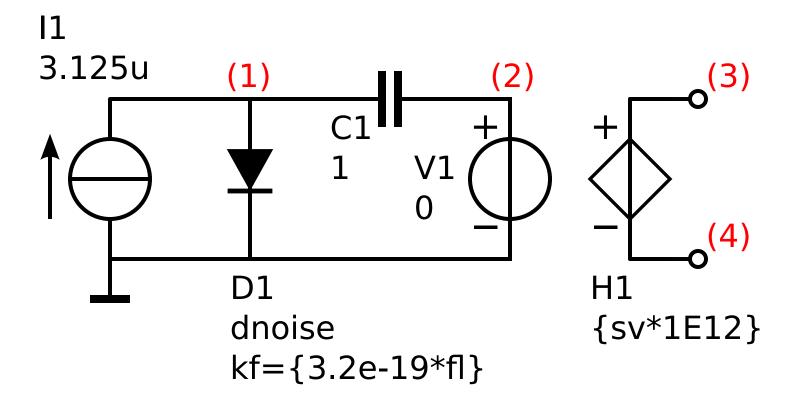
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Spice voltage noise source with floor noise and 1/f noise (from 0.1mHz and up)

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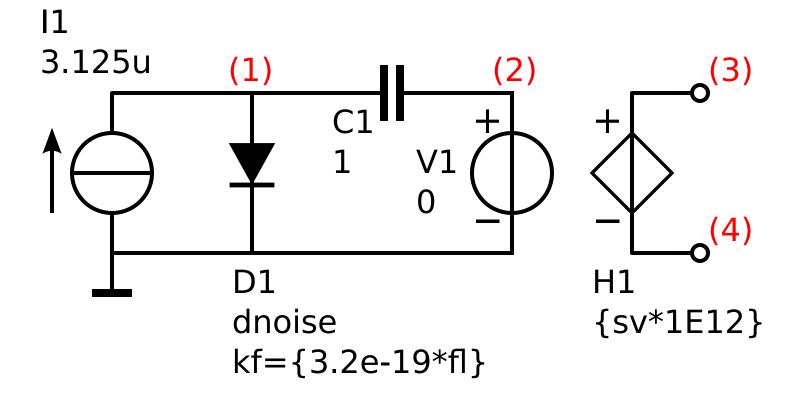


Spice voltage noise source with floor noise and 1/f noise (from 0.1mHz and up)



Diode noise: 
$$S_i = 2qI_D \left(1 + \frac{\mathrm{KF}I_D^{\mathrm{AF}-1}}{2qf}\right)$$

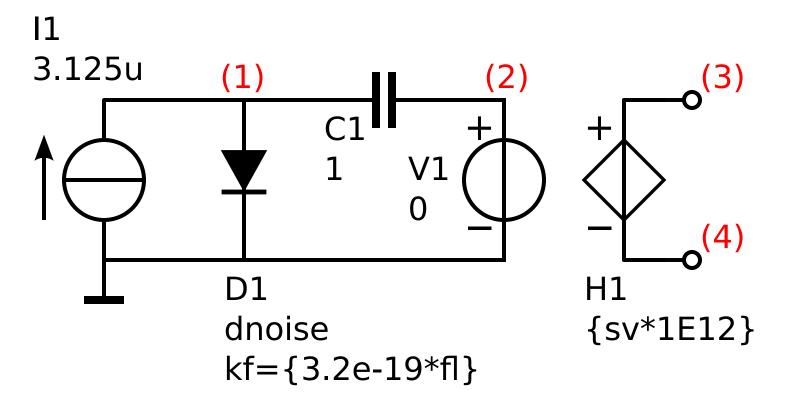
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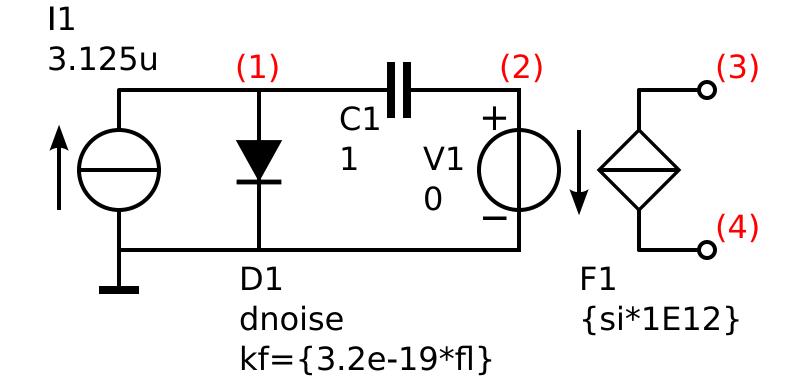
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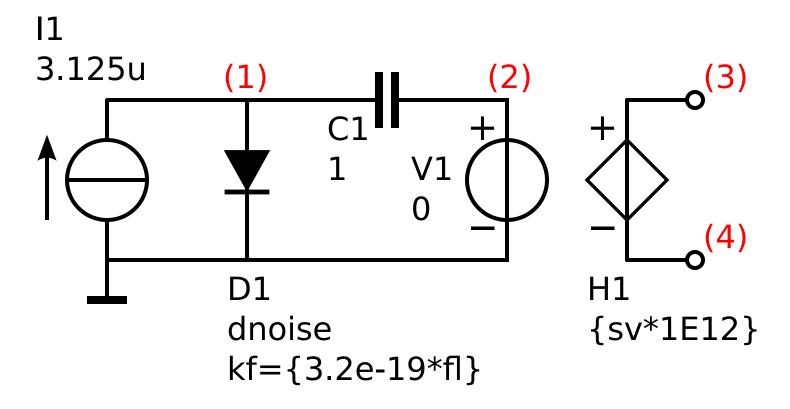


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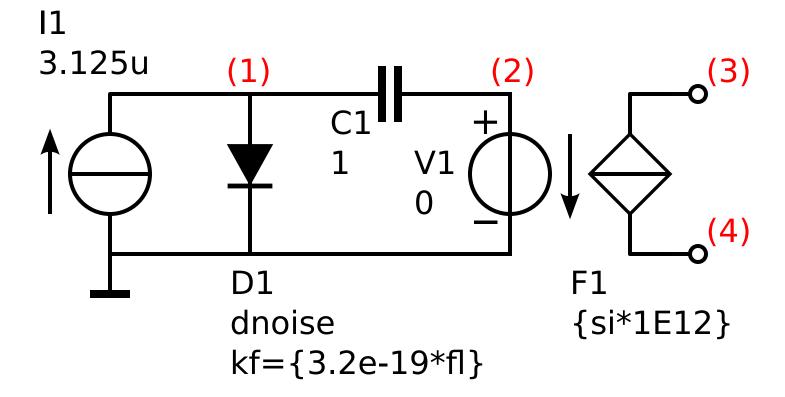


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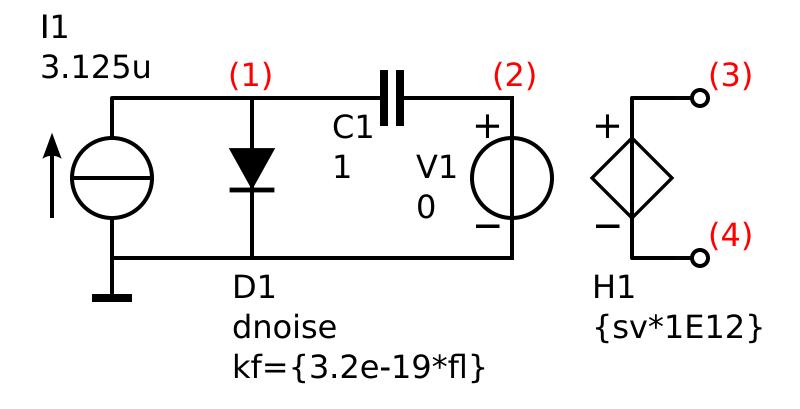
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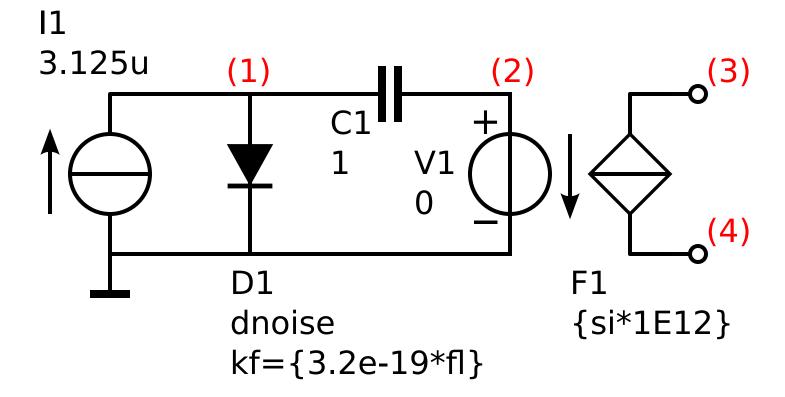
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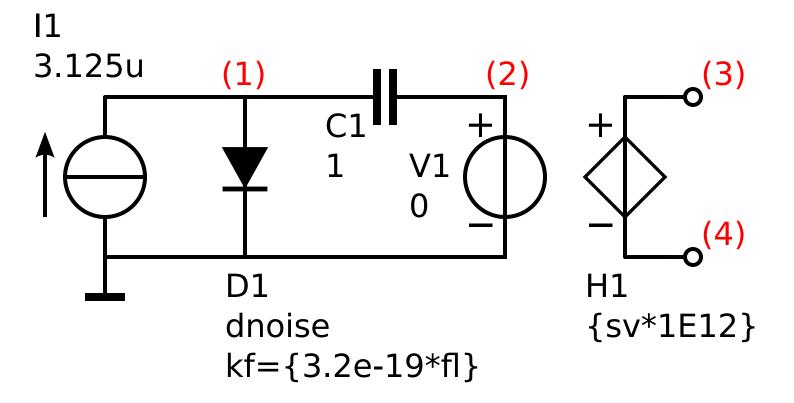
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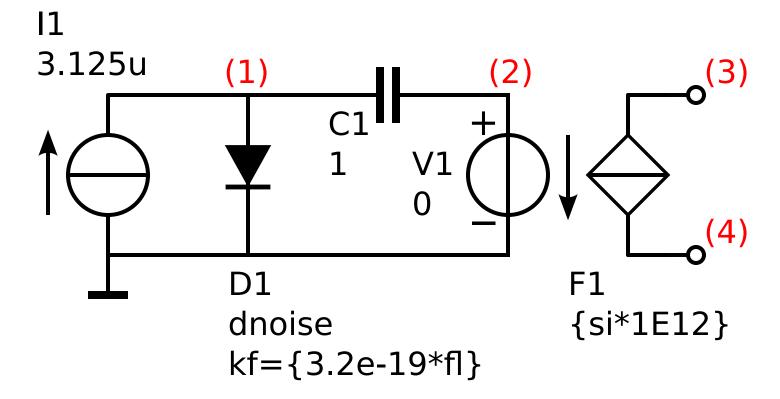
fl: corner frequency 1/f noise

Study: section 8.3.2

Spice voltage noise source with floor noise and 1/f noise (from 0.1mHz and up)



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Study: section 8.3.2

Small-signal dynamic model voltage-feedback OpAmp

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Use active and passive network elements or Laplace blocks

- differential-mode input impedance

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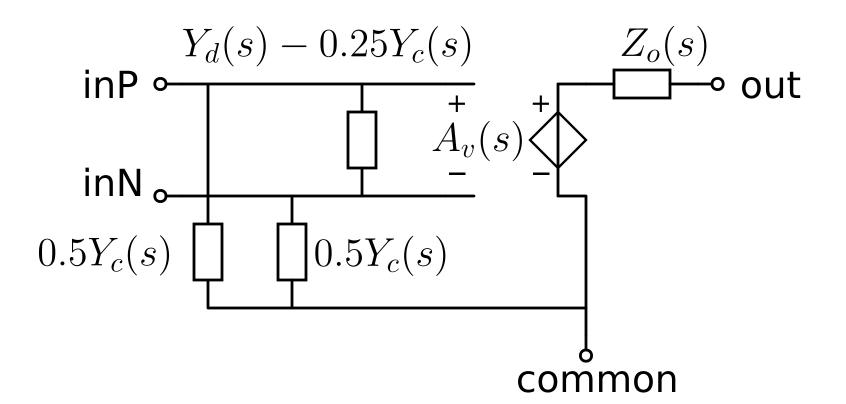
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#### Small-signal dynamic model voltage-feedback OpAmp

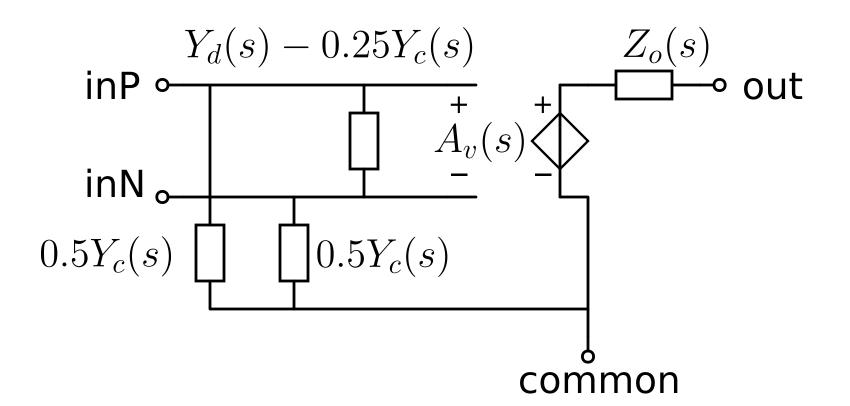
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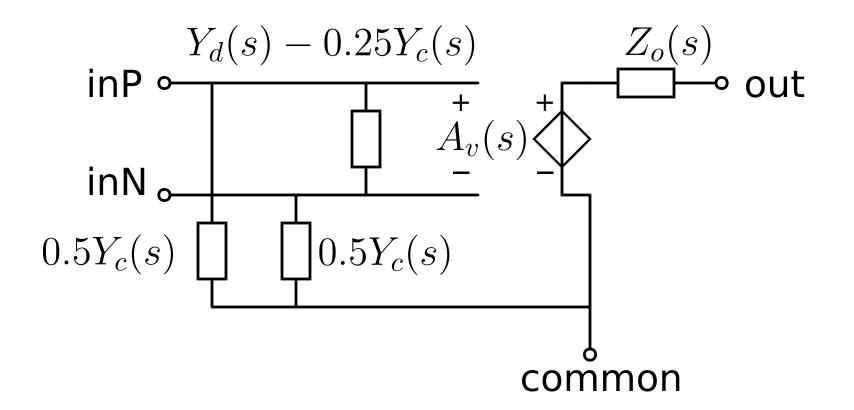
$$Y_d(s) = G_d + sC_d$$



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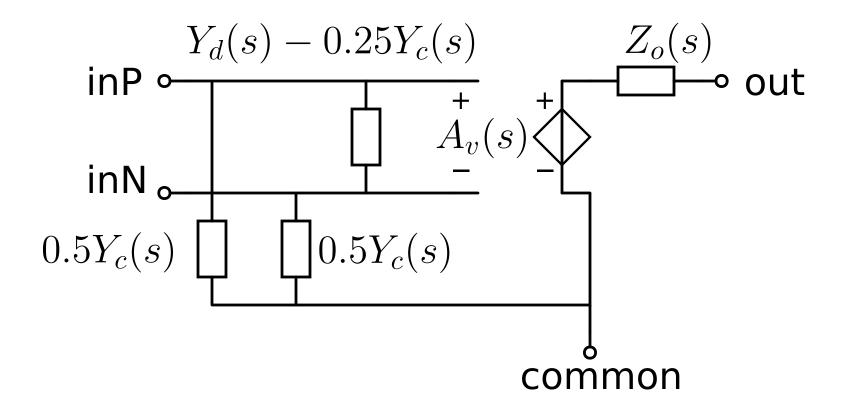
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$$A_v(s) = A_0 \frac{1 + b_1 s + b_2 s^2 + \dots + b_m s^m}{1 + a_1 s + a_2 s^2 + \dots + a_n s^n}$$



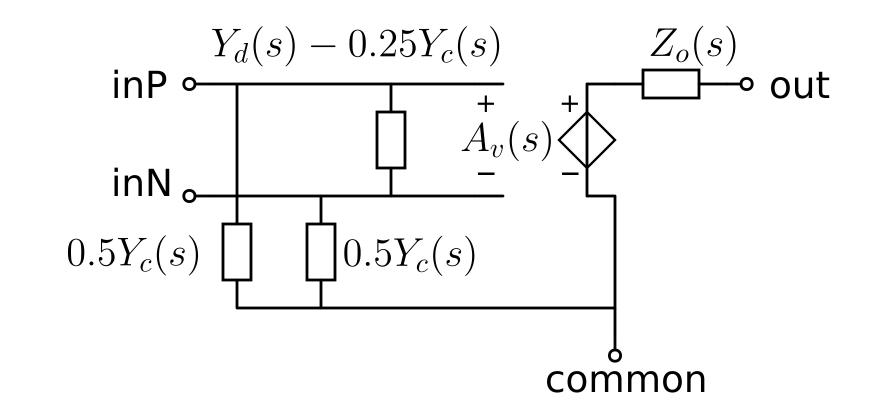
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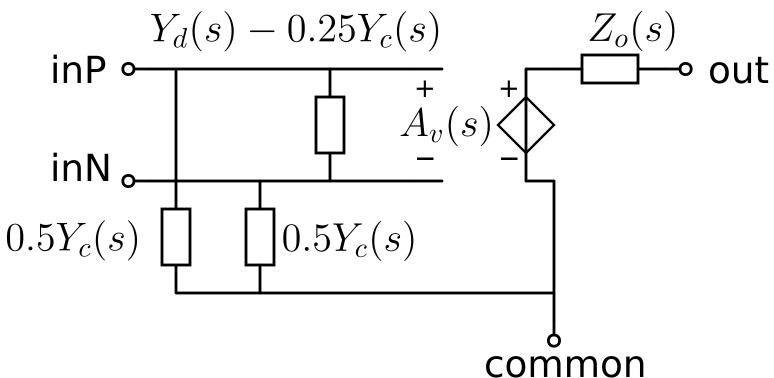
First order: 
$$A_v(s) = A_0 \frac{1}{1 + s \frac{A_0}{2\pi GB}}$$

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Use active and passive network elements or Laplace blocks

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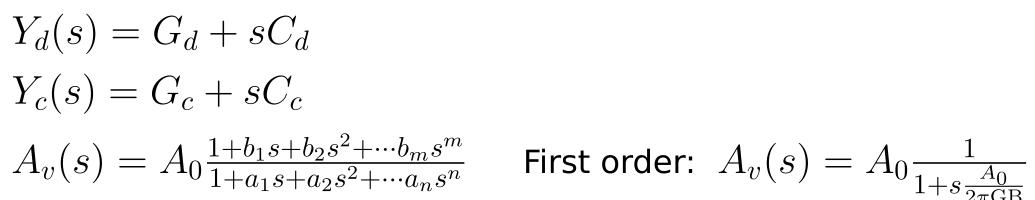
$$\begin{split} Y_d(s) &= G_d + sC_d \\ Y_c(s) &= G_c + sC_c \\ A_v(s) &= A_0 \frac{1 + b_1 s + b_2 s^2 + \cdots b_m s^m}{1 + a_1 s + a_2 s^2 + \cdots a_n s^n} \end{split} \quad \text{First order: } A_v(s) = A_0 \frac{1}{1 + s \frac{A_0}{2\pi GB}} \end{split}$$

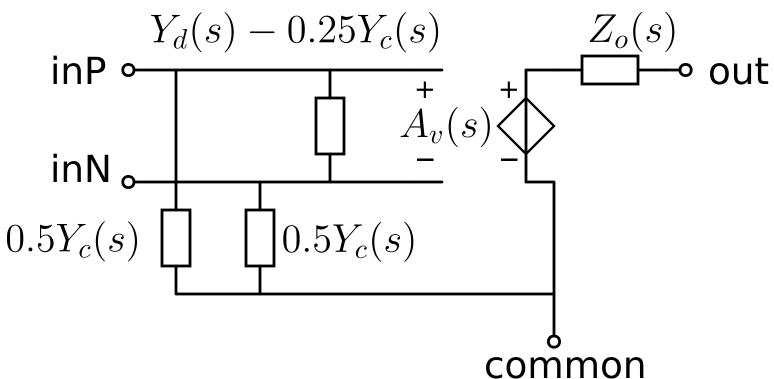


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Small-signal dynamic model current-feedback OpAmp

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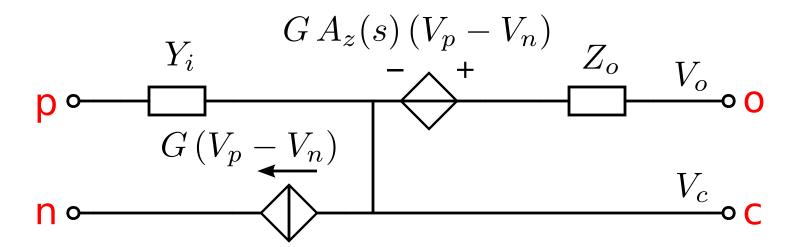
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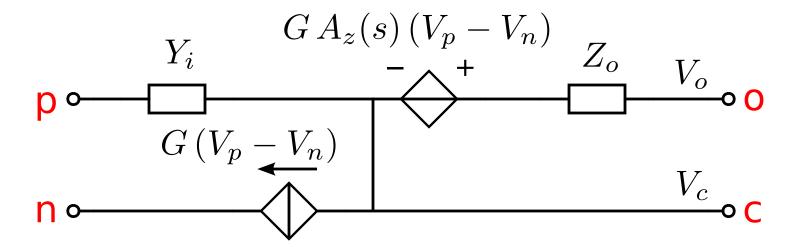
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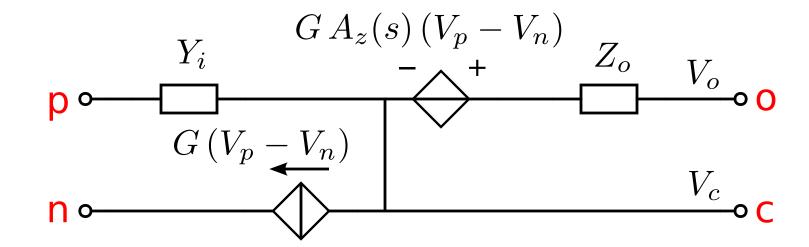
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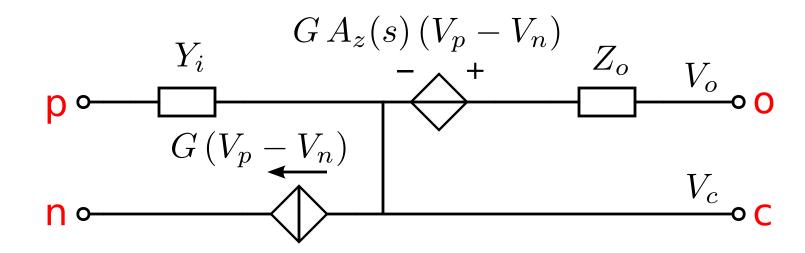


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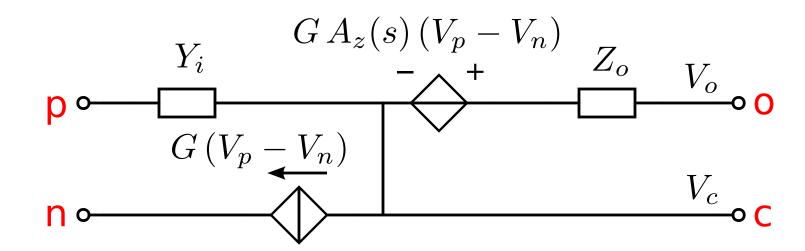
$$Y_i(s) = G_i + sC_i$$

$$A_z(s) = Z_0 \frac{1 + b_1 s + b_2 s^2 + \dots + b_m s^m}{1 + a_1 s + a_2 s^2 + \dots + a_n s^n}$$

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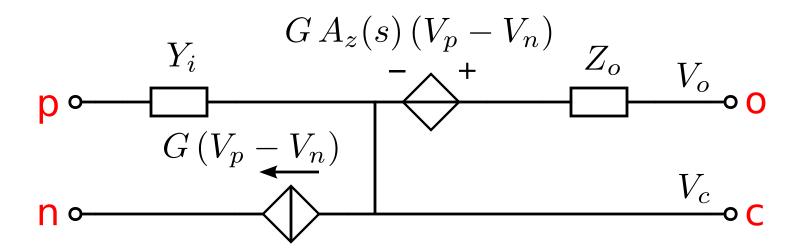
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- transimpedance output stage



$$Y_i(s) = G_i + sC_i$$

$$A_z(s) = Z_0 \frac{1 + b_1 s + b_2 s^2 + \cdots + b_m s^m}{1 + a_1 s + a_2 s^2 + \cdots + a_n s^n} \quad \text{First order: } A_z(s) = R_0 \frac{1}{1 + s \frac{R_0 G}{2\pi GB}}$$